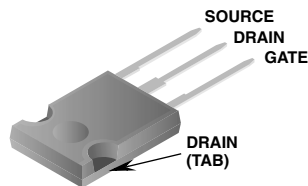


IRFP250N

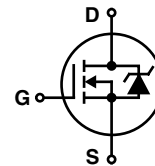
N-Channel Power MOSFET 200V, 30A, 0.075Ω

Features

- Ultra Low On-Resistance
 - $r_{DS(ON)} = 0.052\Omega$ (Typ), $V_{GS} = 10V$
- Simulation Models
 - Temperature Compensated PSpice® and SABER® Electrical Models
 - Spice and SABER® Thermal Impedance Models
- Peak Current vs Pulse Width Curve
- UIS Rating Curve



TO-247



MOSFET Maximum Ratings $T_A = 25^\circ\text{C}$ unless otherwise noted

| Symbol | Parameter | Ratings | Units |
|----------------|---|------------|---------------------|
| V_{DSS} | Drain to Source Voltage | 200 | V |
| V_{GS} | Gate to Source Voltage | ± 20 | V |
| I_D | Drain Current | | |
| | Continuous ($T_C = 25^\circ\text{C}$, $V_{GS} = 10V$) | 30 | A |
| | Continuous ($T_C = 100^\circ\text{C}$, $V_{GS} = 10V$) | 21 | A |
| | Pulsed | Figure 4 | A |
| E_{AS} | Single Pulse Avalanche Energy (Note 1) | 315 | mJ |
| P_D | Power dissipation | 214 | W |
| | Derate above 25°C | 1.4 | W/ $^\circ\text{C}$ |
| T_J, T_{STG} | Operating and Storage Temperature | -55 to 175 | $^\circ\text{C}$ |

Thermal Characteristics

| | | | |
|-----------------|---|------|---------------------------|
| $R_{\theta JC}$ | Thermal Resistance Junction to Case TO-247 | 0.70 | $^\circ\text{C}/\text{W}$ |
| $R_{\theta JA}$ | Thermal Resistance Junction to Ambient TO-247 | 40 | $^\circ\text{C}/\text{W}$ |

Package Marking and Ordering Information

| Device Marking | Device | Package | Reel Size | Tape Width | Quantity |
|----------------|----------|---------|-----------|------------|----------|
| IRFP250N | IRFP250N | TO-247 | Tube | N/A | 30 |

Electrical Characteristics $T_A = 25^\circ\text{C}$ unless otherwise noted

| Symbol | Parameter | Test Conditions | Min | Typ | Max | Units |
|--------|-----------|-----------------|-----|-----|-----|-------|
|--------|-----------|-----------------|-----|-----|-----|-------|

Off Characteristics

| | | | | | | |
|------------|-----------------------------------|---|-----|---|-----------|---------------|
| B_{VDSS} | Drain to Source Breakdown Voltage | $I_D = 250\mu\text{A}$, $V_{GS} = 0\text{V}$ | 200 | - | - | V |
| I_{DSS} | Zero Gate Voltage Drain Current | $V_{DS} = 200\text{V}$, $V_{GS} = 0\text{V}$ | - | - | 25 | μA |
| | | $V_{DS} = 160\text{V}$, $T_C = 150^\circ$ | - | - | 250 | |
| I_{GSS} | Gate to Source Leakage Current | $V_{GS} = \pm 20\text{V}$ | - | - | ± 100 | nA |

On Characteristics

| | | | | | | |
|--------------|----------------------------------|---|----|-------|-------|----------|
| $V_{GS(TH)}$ | Gate to Source Threshold Voltage | $V_{GS} = V_{DS}$, $I_D = 250\mu\text{A}$ | 2 | - | 4 | V |
| $r_{DS(ON)}$ | Drain to Source On Resistance | $I_D = 18\text{A}$, $V_{GS} = 10\text{V}$ | - | 0.052 | 0.075 | Ω |
| g_{fs} | Forward Transconductance | $V_{DS} = 50\text{V}$, $I_D = 18\text{A}$ (Note 2) | 17 | - | - | S |

Dynamic Characteristics

| | | | | | | | |
|--------------|-------------------------------|---|--|------|-----|-----|----|
| C_{ISS} | Input Capacitance | $V_{DS} = 25\text{V}$, $V_{GS} = 0\text{V}$, $f = 1\text{MHz}$ | - | 4023 | - | pF | |
| C_{OSS} | Output Capacitance | | - | 880 | - | pF | |
| C_{RSS} | Reverse Transfer Capacitance | | - | 240 | - | pF | |
| $Q_{g(TOT)}$ | Total Gate Charge at 20V | $V_{GS} = 0\text{V}$ to 20V | $V_{DD} = 160\text{V}$ $I_D = 18\text{A}$ $I_g = 2.0\text{mA}$ | 215 | 280 | nC | |
| $Q_{g(10)}$ | Total Gate Charge at 10V | $V_{GS} = 0\text{V}$ to 10V | | - | 114 | 140 | nC |
| $Q_{g(TH)}$ | Threshold Gate Charge | $V_{GS} = 0\text{V}$ to 2V | | - | 8 | 10 | nC |
| Q_{gs} | Gate to Source Gate Charge | | | - | 14 | - | nC |
| Q_{gd} | Gate to Drain "Miller" Charge | | | - | 44 | - | nC |
| | | | | | | | |

Switching Characteristics ($V_{GS} = 10\text{V}$)

| | | | | | | |
|--------------|---------------------|---|---|----|-----|----|
| t_{ON} | Turn-On Time | $V_{DD} = 100\text{V}$, $I_D = 18\text{A}$ $V_{GS} = 10\text{V}$, $R_{GS} = 3.9\Omega$ | - | - | 69 | ns |
| $t_{d(ON)}$ | Turn-On Delay Time | | - | 16 | - | ns |
| t_r | Rise Time | | - | 30 | - | ns |
| $t_{d(OFF)}$ | Turn-Off Delay Time | | - | 78 | - | ns |
| t_f | Fall Time | | - | 40 | - | ns |
| t_{OFF} | Turn-Off Time | | - | - | 177 | ns |

Drain-Source Diode Characteristics

| | | | | | | |
|----------|-------------------------------|--|---|---|------|----|
| V_{SD} | Source to Drain Diode Voltage | $I_{SD} = 18\text{A}$ | - | - | 1.3 | V |
| t_{rr} | Reverse Recovery Time | $I_{SD} = 18\text{A}$, $dI_{SD}/dt = 100\text{A}/\mu\text{s}$ | - | - | 279 | ns |
| Q_{RR} | Reverse Recovered Charge | $I_{SD} = 18\text{A}$, $dI_{SD}/dt = 100\text{A}/\mu\text{s}$ | - | - | 2000 | nC |

Notes:

- 1: Starting $T_J = 25^\circ\text{C}$, $L = 1.9\text{mH}$, $I_{AS} = 18\text{A}$.
2: Pulse width $\leq 300\mu\text{s}$; duty cycle $\leq 2\%$.

Typical Characteristic

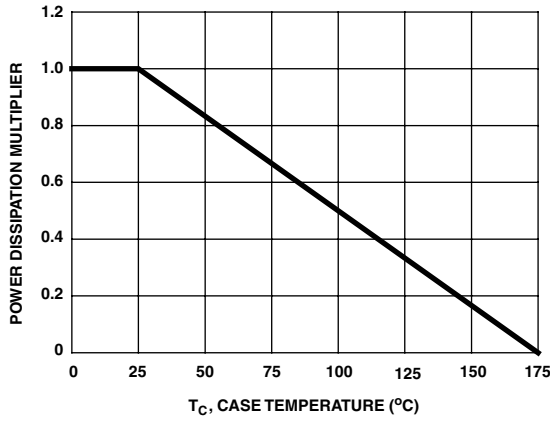


Figure 1. Normalized Power Dissipation vs Ambient Temperature

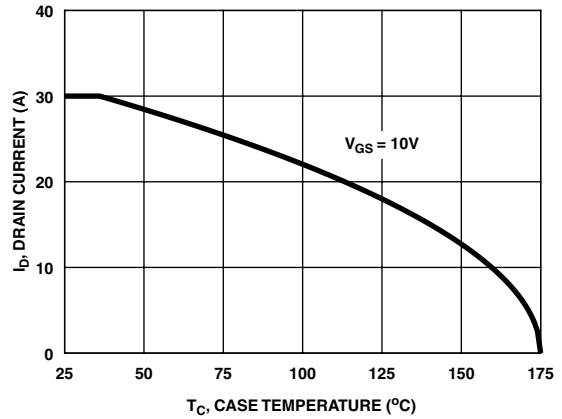


Figure 2. Maximum Continuous Drain Current vs Case Temperature

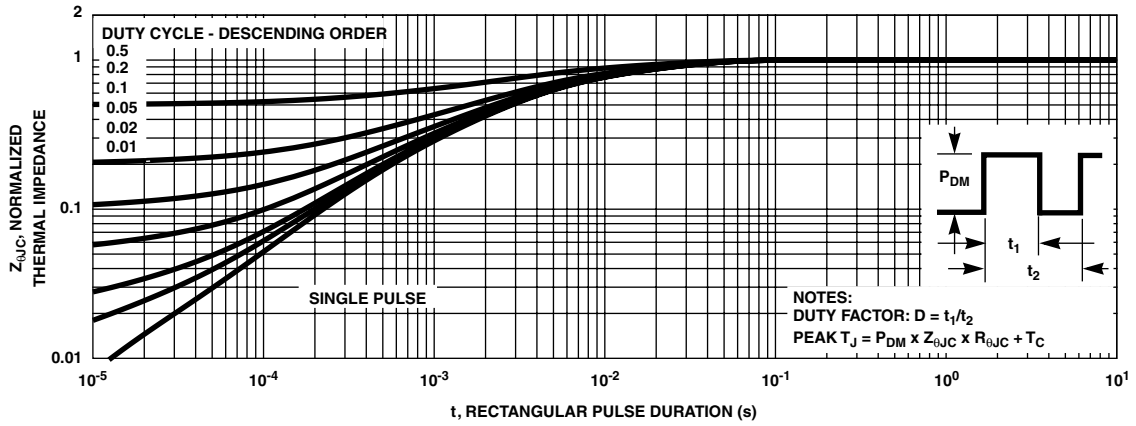


Figure 3. Normalized Maximum Transient Thermal Impedance

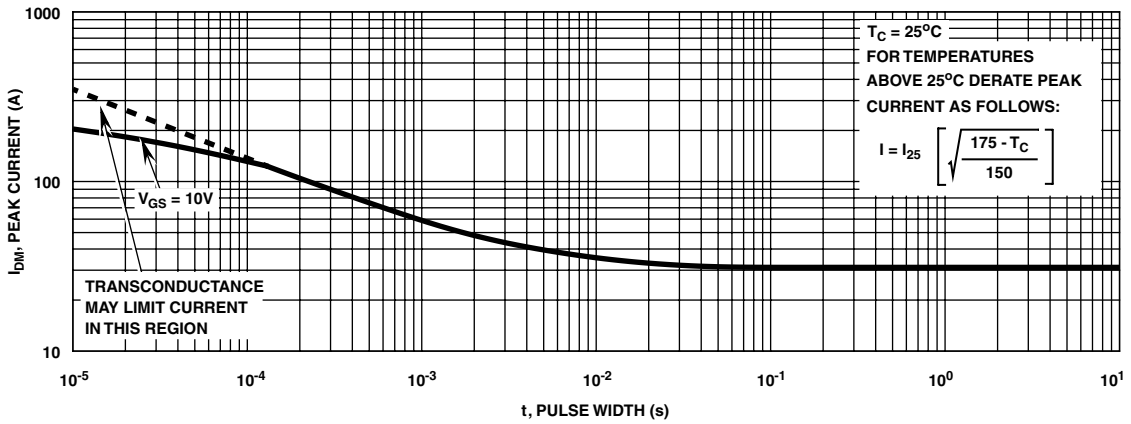


Figure 4. Peak Current Capability

Typical Characteristic (Continued)

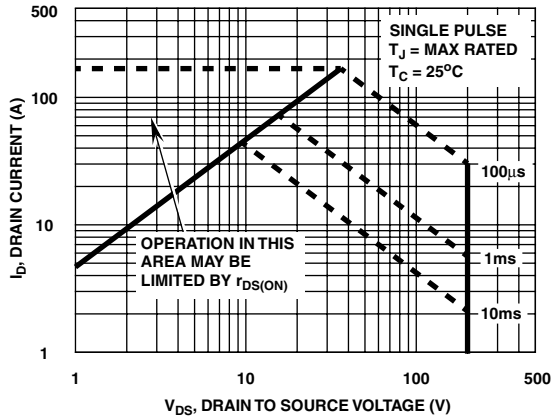


Figure 5. Forward Bias Safe Operating Area

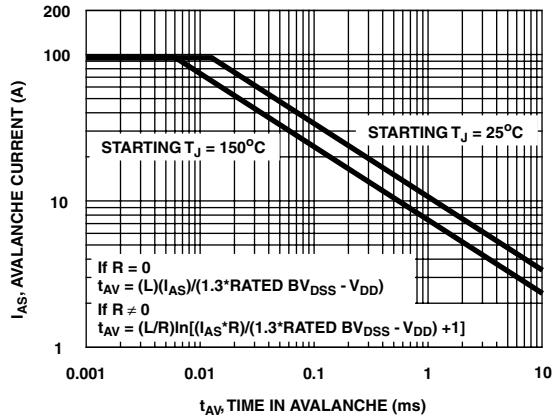


Figure 6. Unclamped Inductive Switching Capability

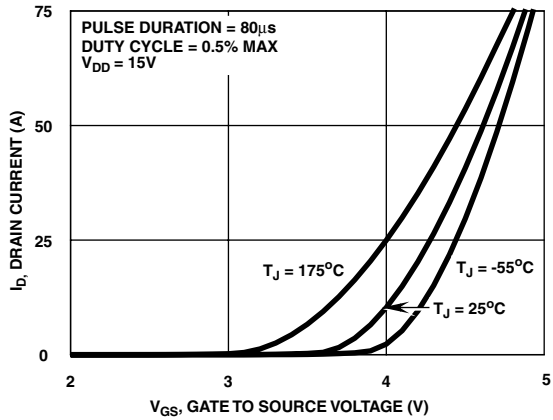


Figure 7. Transfer Characteristics

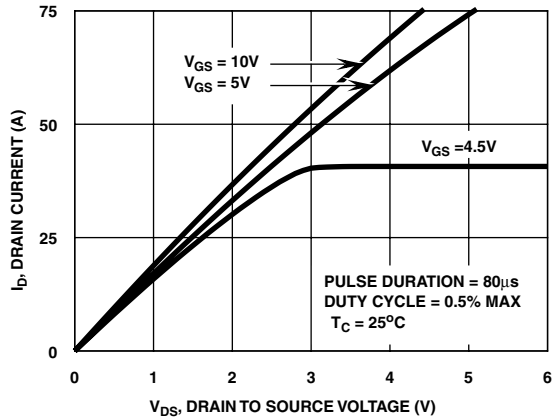


Figure 8. Saturation Characteristics

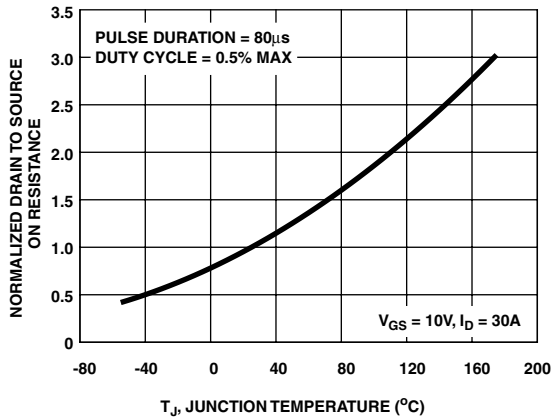


Figure 9. Normalized Drain to Source On Resistance vs Junction Temperature

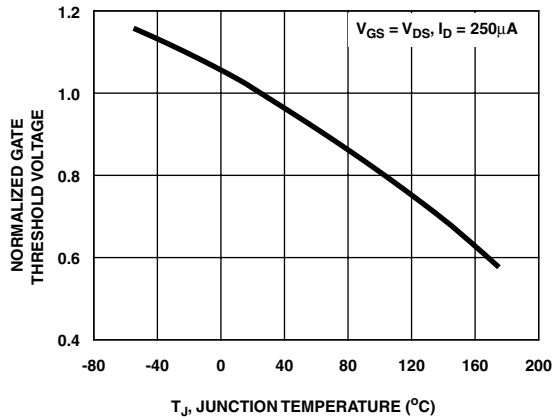


Figure 10. Normalized Gate Threshold Voltage vs Junction Temperature

Typical Characteristic (Continued)

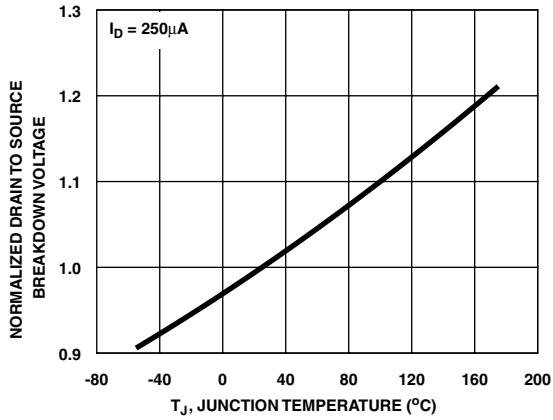


Figure 11. Normalized Drain to Source Breakdown Voltage vs Junction Temperature

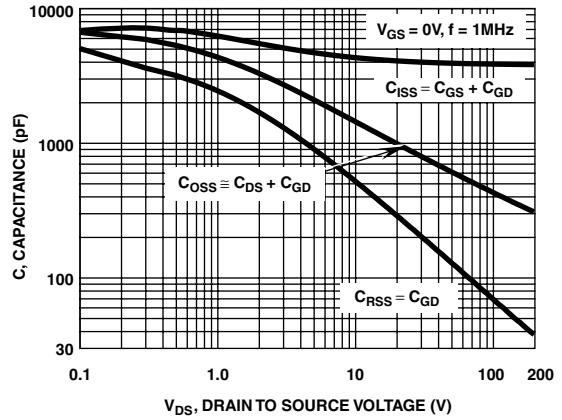


Figure 12. Capacitance vs Drain to Source Voltage

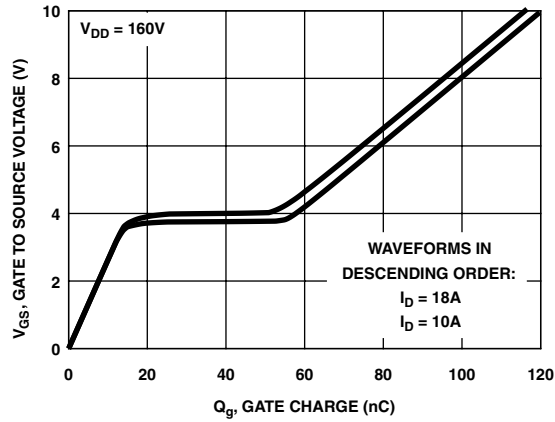


Figure 13. Gate Charge Waveforms for Constant Gate Currents

Test Circuits and Waveforms

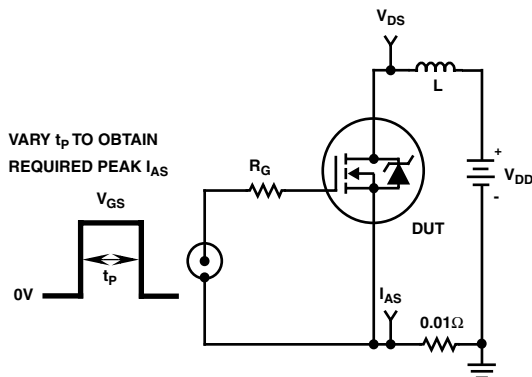


Figure 14. Unclamped Energy Test Circuit

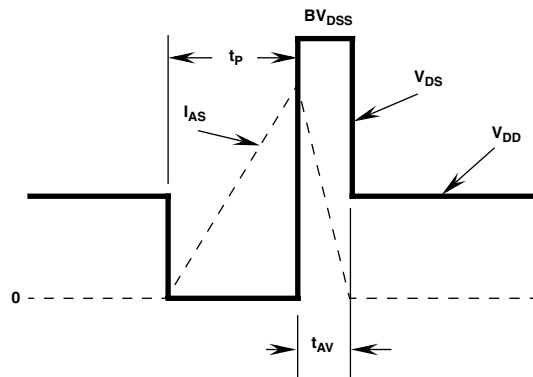


Figure 15. Unclamped Energy Waveforms

Test Circuits and Waveforms (Continued)

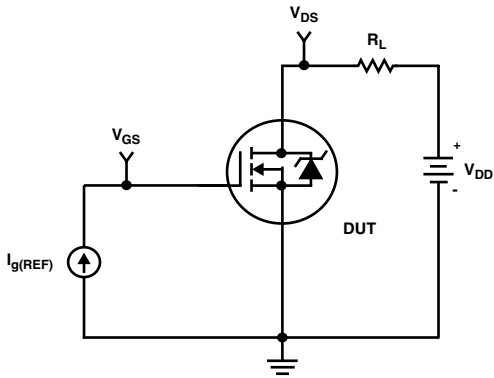


Figure 16. Gate Charge Test Circuit

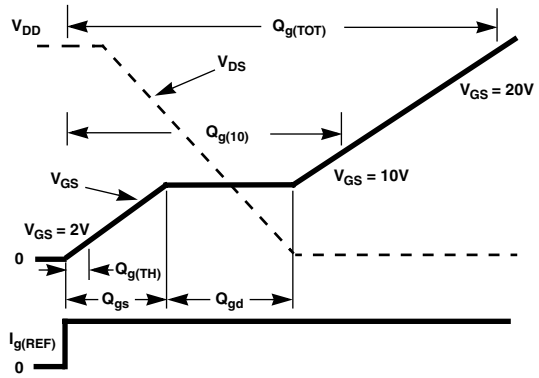


Figure 17. Gate Charge Waveforms

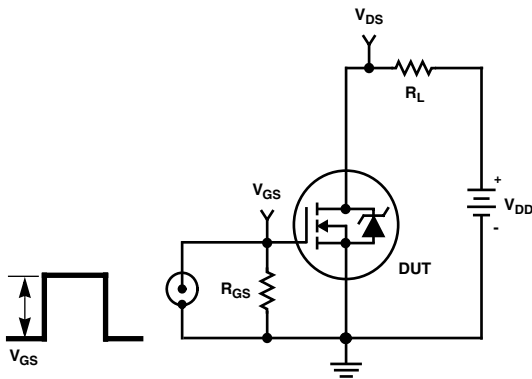


Figure 18. Switching Time Test Circuit

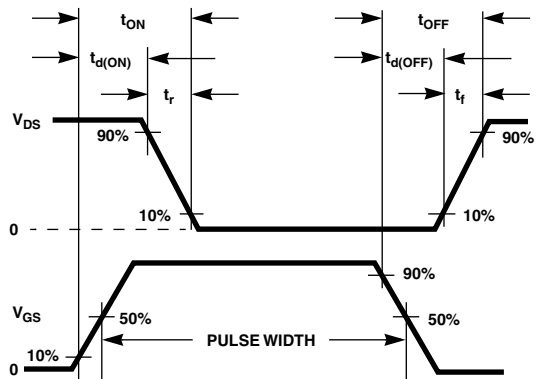


Figure 19. Switching Time Waveforms

PSPICE Electrical Model

.SUBCKT IRFP250N 2 1 3 ; rev May 2001

CA 12 8 6.6e-9
 CB 15 14 6.5e-9
 CIN 6 8 3.80e-9

DBODY 7 5 DBODYMOD
 DBREAK 5 11 DBREAKMOD
 DPLCAP 10 5 DPLCAPMOD

EBREAK 11 7 17 18 221
 EDS 14 8 5 8 1
 EGS 13 8 6 8 1
 ESG 6 10 6 8 1
 EVTHRES 6 21 19 8 1
 EVTEMP 20 6 18 22 1

IT 8 17 1

LDRAIN 2 5 1.0e-9
 LGATE 1 9 8.05e-9
 LSOURCE 3 7 5.8e-9

MMED 16 6 8 8 MMEDMOD
 MSTRO 16 6 8 8 MSTROMOD
 MWEAK 16 21 8 8 MWEAKMOD

RBREAK 17 18 RBREAKMOD 1
 RDRAIN 50 16 RDRAINMOD 5.0e-2
 RGATE 9 20 0.77
 RLDRAIN 2 5 10
 RLGATE 1 9 80.5
 RLSOURCE 3 7 58
 RSLC1 5 51 RSLCMOD 1e-6
 RSLC2 5 50 1e3
 RSOURCE 8 7 RSOURCEMOD 1.8e-3
 RVTHRES 22 8 RVTHRESMOD 1
 RVTEMP 18 19 RVTEMPMOD 1

S1A 6 12 13 8 S1AMOD
 S1B 13 12 13 8 S1BMOD
 S2A 6 15 14 13 S2AMOD
 S2B 13 15 14 13 S2BMOD

VBAT 22 19 DC 1

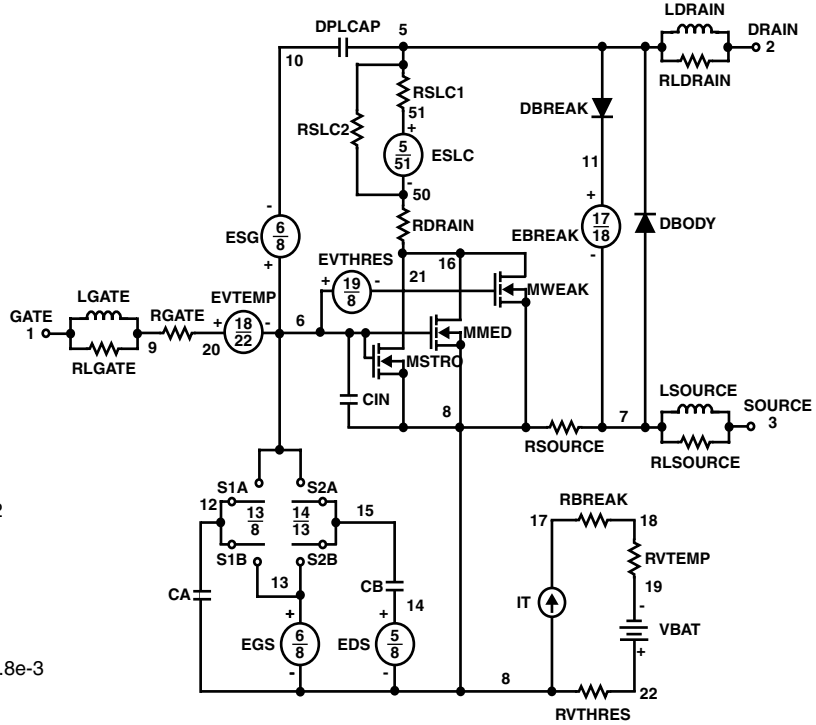
ESLC 51 50 VALUE={(V(5,51)/ABS(V(5,51)))*(PWR(V(5,51))/(1e-6*100),2.5)}

.MODEL DBODYMOD D (IS = 2.8e-12 RS = 3.0e-3 XTI = 5.5 TRS1 = 3.5e-3 TRS2 = 1e-5 CJO = 2.55e-9 TT = 1.52e-7 M = 0.42)
 .MODEL DBREAKMOD D (RS = 1.2e-0 TRS1 = 1e-3 TRS2 = 1e-6)
 .MODEL DPLCAPMOD D (CJO = 4.6e-9 IS = 1e-30 N = 10 M = 0.9)
 .MODEL MMEDMOD NMOS (VTO = 3.05 KP = 2.5 IS = 1e-30 N = 10 TOX = 1 L = 1u W = 1u RG = 0.77)
 .MODEL MSTROMOD NMOS (VTO = 3.55 KP = 100 IS = 1e-30 N = 10 TOX = 1 L = 1u W = 1u)
 .MODEL MWEAKMOD NMOS (VTO = 2.69 KP = 0.05 IS = 1e-30 N = 10 TOX = 1 L = 1u W = 1u RG = 7.70)
 .MODEL RBREAKMOD RES (TC1 = 1.27e-3 TC2 = 1.0e-6)
 .MODEL RDRAINMOD RES (TC1 = 9.90e-3 TC2 = 3.60e-5)
 .MODEL RSLCMOD RES (TC1 = 3.0e-3 TC2 = 1.0e-6)
 .MODEL RSOURCEMOD RES (TC1 = 1e-3 TC2 = 1e-6)
 .MODEL RVTHRESMOD RES (TC1 = -2.90e-3 TC2 = -1.10e-5)
 .MODEL RVTEMPMOD RES (TC1 = -2.80e-3 TC2 = 1.70e-7)

.MODEL S1AMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = -5.5 VOFF = -4.5)
 .MODEL S1BMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = -4.5 VOFF = -5.5)
 .MODEL S2AMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = -0.3 VOFF = 0.4)
 .MODEL S2BMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = 0.4 VOFF = -0.3)

.ENDS

NOTE: For further discussion of the PSPICE model, consult **A New PSPICE Sub-Circuit for the Power MOSFET Featuring Global Temperature Options**; IEEE Power Electronics Specialist Conference Records, 1991, written by William J. Hepp and C. Frank Wheatley.



SABER Electrical Model

REV May 2001

template IRFP250N n2,n1,n3

electrical n2,n1,n3

{

var i iscl

dp..model dbodymod = (isl = 2.8e-12, rs = 3.0e-3, xti = 5.5, trs1 = 3.5e-3, trs2 = 1.0e-5, cjo = 2.55e-9, tt = 1.52e-7, m = 0.42)

dp..model dbreakmod = (rs = 1.2, trs1 = 1.0e-3, trs2 = 1.0e-6)

dp..model dplcapmod = (cjo = 4.6e-9, isl = 10e-30, nl=10, m = 0.9)

m..model mmedmod = (type=_n, vto = 3.05, kp = 2.5, is = 1e-30, tox = 1)

m..model mstrongmod = (type=_n, vto = 3.55, kp = 100, is = 1e-30, tox = 1)

m..model mweakmod = (type=_n, vto = 2.69, kp = 0.05, is = 1e-30, tox = 1, rs=0.1)

sw_vcsp..model s1amod = (ron = 1e-5, roff = 0.1, von = -5.5, voff = -4.5)

sw_vcsp..model s1bmod = (ron = 1e-5, roff = 0.1, von = -4.5, voff = -5.5)

sw_vcsp..model s2amod = (ron = 1e-5, roff = 0.1, von = -0.3, voff = 0.4)

sw_vcsp..model s2bmod = (ron = 1e-5, roff = 0.1, von = 0.4, voff = -0.3)

c.ca n12 n8 = 6.6e-9

c.cb n15 n14 = 6.5e-9

c.cin n6 n8 = 3.8e-9

dp.dbody n7 n5 = model=dbodymod

dp.dbreak n5 n11 = model=dbreakmod

dp.dplcap n10 n5 = model=dplcapmod

i.it n8 n17 = 1

l.ldrain n2 n5 = 1.00e-9

l.lgate n1 n9 = 8.05e-9

l.lsource n3 n7 = 5.80e-9

m.mmed n16 n6 n8 n8 = model=mmedmod, l=1u, w=1u

m.mstrong n16 n6 n8 n8 = model=mstrongmod, l=1u, w=1u

m.mweak n16 n21 n8 n8 = model=mweakmod, l=1u, w=1u

res.rbreak n17 n18 = 1, tc1 = 1.27e-3, tc2 = 1.00e-6

res.rdrain n50 n16 = 5.0e-2, tc1 = 9.9e-3, tc2 = 3.6e-5

res.rgate n9 n20 = 0.77

res.rldrain n2 n5 = 10

res.rlgate n1 n9 = 80.5

res.rlsource n3 n7 = 58

res.rslc1 n5 n51 = 1e-6, tc1 = 3e-3, tc2 = -1.0e-6

res.rslc2 n5 n50 = 1e3

res.rsource n8 n7 = 1.8e-3, tc1 = 1.0e-3, tc2 = 1e-6

res.rvtemp n18 n19 = 1, tc1 = -2.8e-3, tc2 = 1.70e-6

res.rvthres n22 n8 = 1, tc1 = -2.9e-3, tc2 = 1.1e-5

spe.ebreak n11 n7 n17 n18 = 221

spe.eds n14 n8 n5 n8 = 1

spe.egs n13 n8 n6 n8 = 1

spe.esg n6 n10 n6 n8 = 1

spe.evtemp n20 n6 n18 n22 = 1

spe.evthres n6 n21 n19 n8 = 1

sw_vcsp.s1a n6 n12 n13 n8 = model=s1amod

sw_vcsp.s1b n13 n12 n13 n8 = model=s1bmod

sw_vcsp.s2a n6 n15 n14 n13 = model=s2amod

sw_vcsp.s2b n13 n15 n14 n13 = model=s2bmod

v.vbat n22 n19 = dc=1

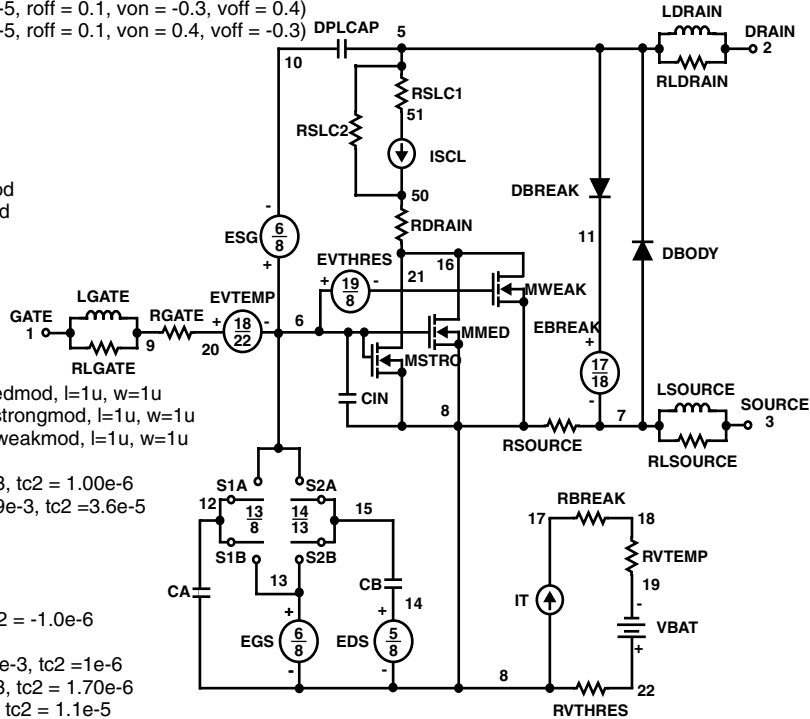
equations {

i (n51->n50) +=iscl

iscl: v(n51,n50) = ((v(n5,n51))/(1e-9+abs(v(n5,n51))))*((abs(v(n5,n51))*1e6*100)** 2.5)

}

}



SPICE Thermal Model

REV May 2001

IRFP250N

CTHERM1 th 6 6.45e-3
 CTHERM2 6 5 3.00e-2
 CTHERM3 5 4 1.40e-2
 CTHERM4 4 3 1.65e-2
 CTHERM5 3 2 4.85e-2
 CTHERM6 2 tl 1.00e-1

RTHERM1 th 6 3.24e-3
 RTHERM2 6 5 8.08e-3
 RTHERM3 5 4 2.28e-2
 RTHERM4 4 3 1.00e-1
 RTHERM5 3 2 1.10e-1
 RTHERM6 2 tl 1.40e-1

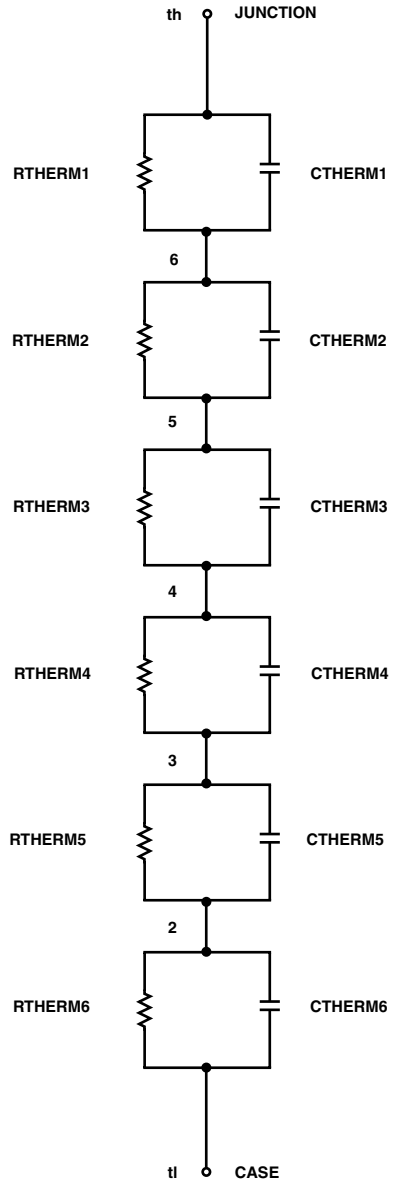
SABER Thermal Model

SABER thermal model IRFP250N

```

template thermal_model th tl
thermal_c th, tl
{
    ctherm.ctherm1 th 6 = 6.45e-3
    ctherm.ctherm2 6 5 = 3.00e-2
    ctherm.ctherm3 5 4 = 1.40e-2
    ctherm.ctherm4 4 3 = 1.65e-2
    ctherm.ctherm5 3 2 = 4.85e-2
    ctherm.ctherm6 2 tl = 1.00e-1

    rtherm.rtherm1 th 6 = 3.24e-3
    rtherm.rtherm2 6 5 = 8.08e-3
    rtherm.rtherm3 5 4 = 2.28e-2
    rtherm.rtherm4 4 3 = 1.00e-1
    rtherm.rtherm5 3 2 = 1.10e-1
    rtherm.rtherm6 2 tl = 1.40e-1
}
    
```



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| Bottomless TM | FAST _r TM | OPTOPLANAR TM | STAR*POWER TM | |
| CoolFET TM | FRFET TM | PACMAN TM | Stealth TM | |
| CROSSVOLT TM | GlobalOptoisolator TM | POP TM | SuperSOT TM -3 | |
| DenseTrench TM | GTO TM | Power247 TM | SuperSOT TM -6 | |
| DOMET TM | HiSeC TM | PowerTrench [®] | SuperSOT TM -8 | |
| EcoSPARK TM | ISOPLANAR TM | QFET TM | SyncFET TM | |
| E ² CMOS TM | LittleFET TM | QST TM | TinyLogic TM | |
| EnSigna TM | MicroFET TM | QT Optoelectronics TM | TruTranslation TM | |
| FACT TM | MicroPak TM | Quiet Series TM | UHC TM | |
| FACT Quiet Series TM | MICROWIRE TM | SILENT SWITCHER [®] | UltraFET [®] | |

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2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

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| Datasheet Identification | Product Status | Definition |
|--------------------------|------------------------|---|
| Advance Information | Formative or In Design | This datasheet contains the design specifications for product development. Specifications may change in any manner without notice. |
| Preliminary | First Production | This datasheet contains preliminary data, and supplementary data will be published at a later date. Fairchild Semiconductor reserves the right to make changes at any time without notice in order to improve design. |
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